On the use of the Cauchy integral formula

- Motivation
- Cauchy & Biot and Savart
- Surface current method
- Example
 - Dipole
 - Quadrupole
 - undulator



Augustin-Louis Cauchy 1789-1857



INTRODUCTORY REMARK: 18TH -19TH CENTURY



FIELD INTEGRAL MEASUREMENTS

Potential difficulty with measurement in apertures with large aspect ratio Example with usual circular measurement

Circular bore radius ? 2 ? Matching the ROI to the aperture OK



THE TOOL: CAUCHY INTEGRAL FORMULA

 $b(z) = b_y(z) + ib_x(z)$ analytic in domain D simply connected and C an oriented closed path in D

$$b(z) = \frac{1}{2\pi i} \int_{C} \frac{b(z_c)}{z_c - z} dz_c$$

Cauchy integral formula
In other words, if we know $b(z)$ on the contour *C*, $b(z)$ can be reconstructed at any point inside *C*.
Also:
$$\int_{C} b(z) dz = 0$$
 (Cauchy' theorem) (Cauchy' theorem)



CAUCHY INTEGRAL & BIOT AND SAVART

Some little transformation & interpretation

$$b(z) = \frac{1}{2\pi i} \int_{C} \frac{b(z_c)}{z_c - z} dz_c$$

$$b(z) = \frac{1}{2\pi} \int_{C} \frac{b(z_c)i \operatorname{Sign}(dz_c)}{z - z_c} |dz_c| \qquad Sign(z) = \frac{z}{|z|}$$



Ζ

$$b n = (b_y + ib_x) (n_x + in_y)$$

= $j_s + i\sigma_s = k_s$

$$b(z) = \frac{1}{2\pi} \int_{C} \frac{k_s}{z - z_c} |dz_c| \qquad \text{2D Biot & Savart law}$$

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CONSTANT SURFACE DISTRIBUTION



Potential
$$\mathcal{A}(z, z_1, z_2) = a + iv = k_s (u_2 - u_1 - u_2 \ln(u_2) + u_1 \ln(u_1))$$
 Sign(z) =

Field
$$\mathcal{B}(z, z_1, z_2) = b_y + ib_x = \frac{k_s}{Sign(z_2 - z_1)} \ln \frac{u_1}{u_2}$$
 $u_1 = \frac{z - z_1}{Sign(z_2 - z_1)}$

Order n gradient
1= quadrupole term
$$\mathcal{G}(z, z_1, z_2, n) = \frac{k_s(n-1)! \ (-1)^{n-1}}{Sign(z_2 - z_1)} \left(\frac{1}{(z-z_1)^n} - \frac{1}{(z-z_2)^n}\right) \qquad u_2 = \frac{z-z_2}{Sign(z_2 - z_1)}$$



 $\frac{z}{|z|}$

STRETCHED WIRE

Wire moving from A to B



$$\phi_{AB} = \int_{A}^{B} b_{y} dx = \int_{A}^{B} \frac{\partial a}{\partial x} dx = a(B) - a(A)$$

We only deal the real part of the complex potential $\ensuremath{\mathcal{A}}$

Closed boundary with N segments



Potential a known at the middle of each segment using stretched wire measurements.

Focus: determine a surface current distribution on the boundary generating the potential a



SURFACE CURRENT AT BOUNDARY

 $A = M J_S$ A vector with potential $a_i, i = (1, N)$

 J_s vector with surface current J_{si} , i = (1, N)

M square $N \times N$ matrix with terms

$$m_{ij} = Re(\mathcal{A}\left((z_i + z_{i+1})/2, z_j, z_{j+1}\right)$$

Segment i delimited by z_i and z_{i+1}

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M is a (symmetric) matrix which can be inverted

$$J_s = M^{-1}A$$





FIELD CALCULATION

With j_s known on boundary C

field
$$b(z) = \sum_{i=1}^{N} j_{si} \mathcal{B}(z, z_i, z_{i+1})$$
 potential $a(z) = \sum_{i=1}^{N} j_{si} \mathcal{A}(z, z_i, z_{i+1})$

gradient $g(z,n) = \sum_{i=1}^{N} j_{si} \mathcal{G}(z, z_i, z_{i+1}, n)$

For any point inside boundary

Outside boundary: field is not diverging but do not represent correctly the magnet



APPLICATION TO MAGNETIC MEASUREMENTS



- Synchronized Newport axes
- XPS controller
- Carbon fiber wire
- Keythley 2182 nano-voltmeter
- Wire moved along a **closed** boundary
- Constant speed on trajectory ~ 20 mm/s
- Voltage integrated over each segment





- 332 segments
- pole gap 24.5 mm
- Segment length 0.7 mm



DLS (DIPOLE)





SOME COOL STUFF



Looking at pole details ~ 1.25 mm from pole



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HARMONIC ANALYSIS

Can be done directly at any point inside contour





EXAMPLE2: QUADRUPOLE



EXAMPLE 2: QUADRUPOLE



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EXAMPLE 3: UNDULATOR

Undulator gap 100 mm 400 segments





EXAMPLE 3: UNDULATOR



At 2 mm from magnet surface









DEVELOPMENTS

• Efficient path descriptor



- Additional building elements
 - Portion of arcs (analytical formulation done)
 - Polynomial surface distribution



- A new method for integrated field measurement developed
- Well adapted for stretched wire
- Method was used for the measurement of all DLs
- Presently in routine use for ID field integral measurements @ ESRF
 - Seems interesting for small gap devices



THANK YOU





